

# EK 307: Electric Circuits

Fall 2017

Lecture 18

Nov 9, 2017

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# Administrivia

- Homework 7 due
- Midterm #2: Thu Nov 16

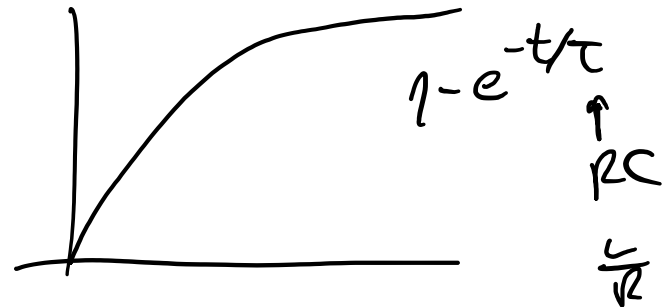
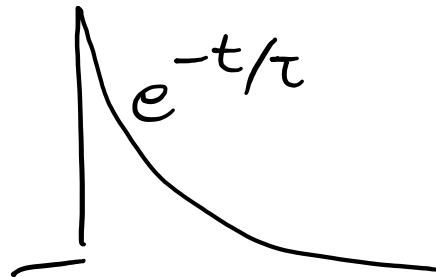
# Last time (Lecture 17):

1. Driven RC and RL circuits
2. First-order op-amp circuits

$$i = C \frac{dv}{dt}$$

- ① initial  $v_c(0)$  or  $i_L(0)$
- ② st  $v_c(\infty)$ ,  $i_L(\infty)$
- ③ RC.

(Chapter 7)



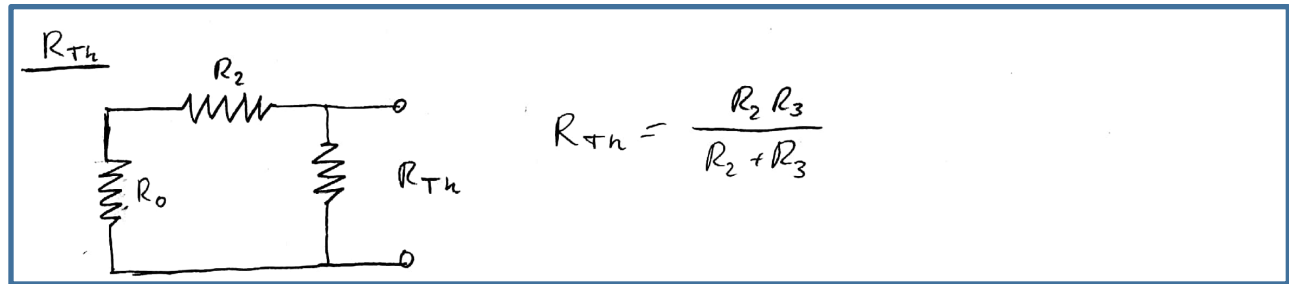
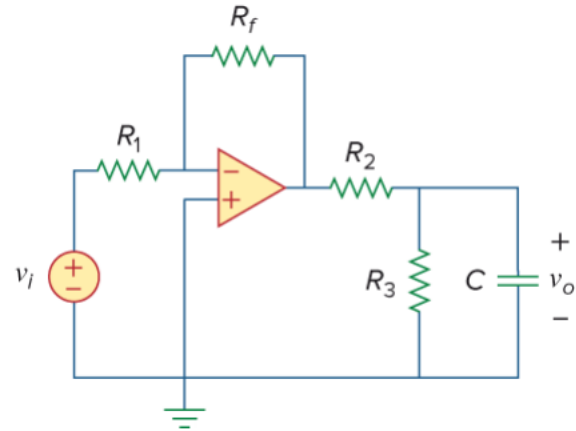
# Lecture 18: Chapter 8 – 2<sup>nd</sup> order ccts

1. Driven RC and RL circuits
2. First-order op-amp circuits

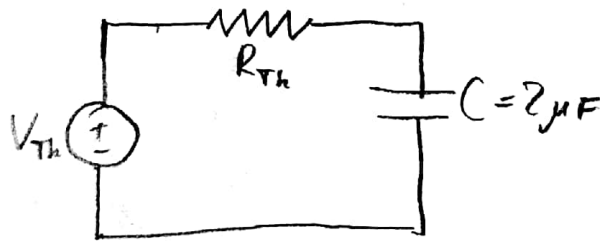
**(covering sections 7.5-7.7)**

## Example 7.16

Find the step response  $v_o(t)$  for  $t > 0$  in the op amp circuit of **Fig. 7.59**. Let  $v_i = 2u(t)$  V,  $R_1 = 20 \text{ k}\Omega$ ,  $R_f = 50 \text{ k}\Omega$ ,  $R_2 = R_3 = 10 \text{ k}\Omega$ ,  $C = 2 \mu\text{F}$ .



equivalent circuit



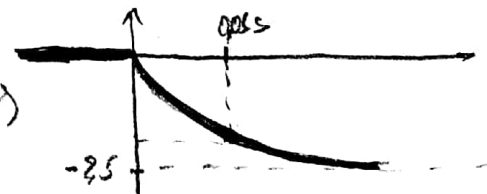
$$V_{Th} = -\frac{10}{20} \cdot \frac{50}{20} \cdot 2u(t) = -2,5u(t)$$

$$R_{Th} = \frac{R_2 R_3}{R_2 + R_3} = \frac{10 \cdot 10}{20} \text{ k}\Omega = 5 \text{ k}\Omega$$

$$\tau = R_{Th} \cdot C = 5 \cdot 10^3 \Omega \cdot 2 \cdot 10^{-6} \text{ F} = 0,01 \text{ s}$$

$$v_o(t) = v_o(\infty) + [v_o(0) - v_o(\infty)] e^{-\frac{t}{\tau}}, \quad t > 0$$

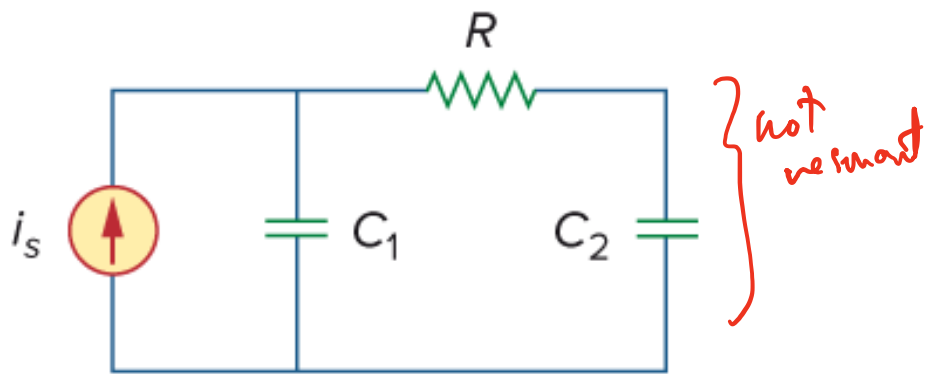
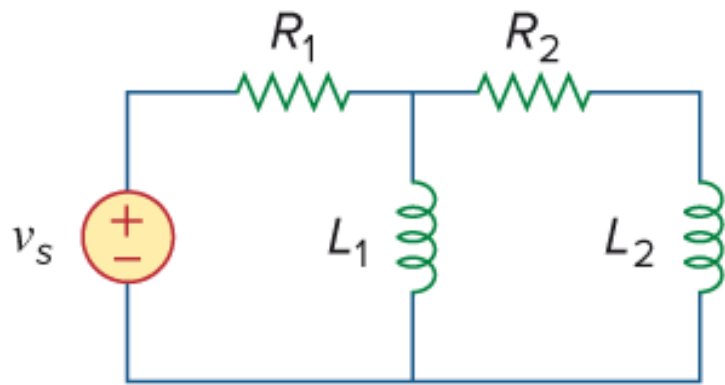
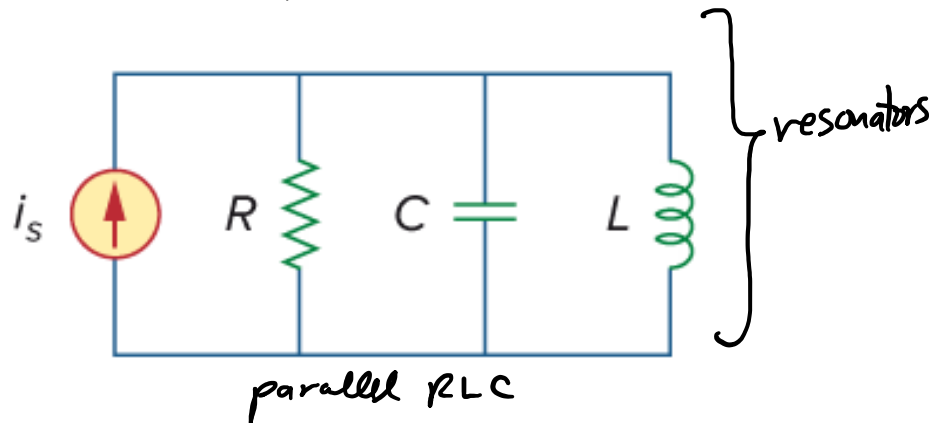
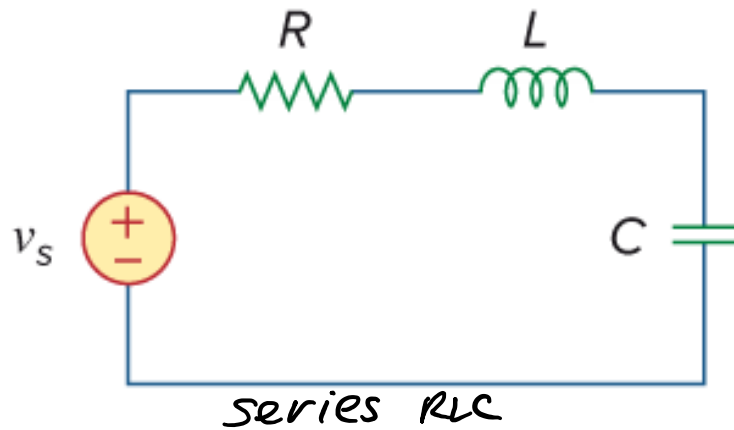
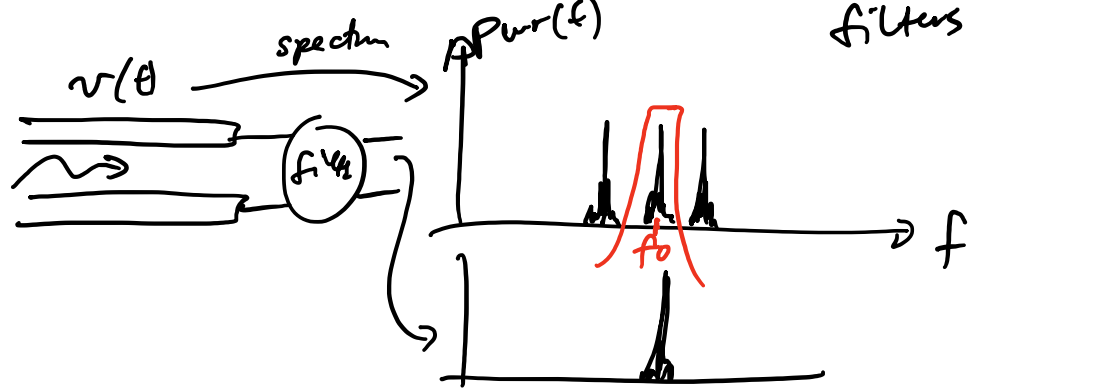
$$v_o(t) = -2,5 + [0 + 2,5] e^{-\frac{t}{0,01 \text{ s}}} = -2,5(1 - e^{-\frac{t}{0,01 \text{ s}}}) u(t)$$



# Second-order circuits

- 1<sup>st</sup>-order circuits  $\Rightarrow$  1<sup>st</sup> order diff eq.  $\Rightarrow$  1 energy storage element
- 2<sup>nd</sup>-order circuits  $\Rightarrow$  2<sup>nd</sup> order diff eq.  $\Rightarrow$  2 energy storage elements

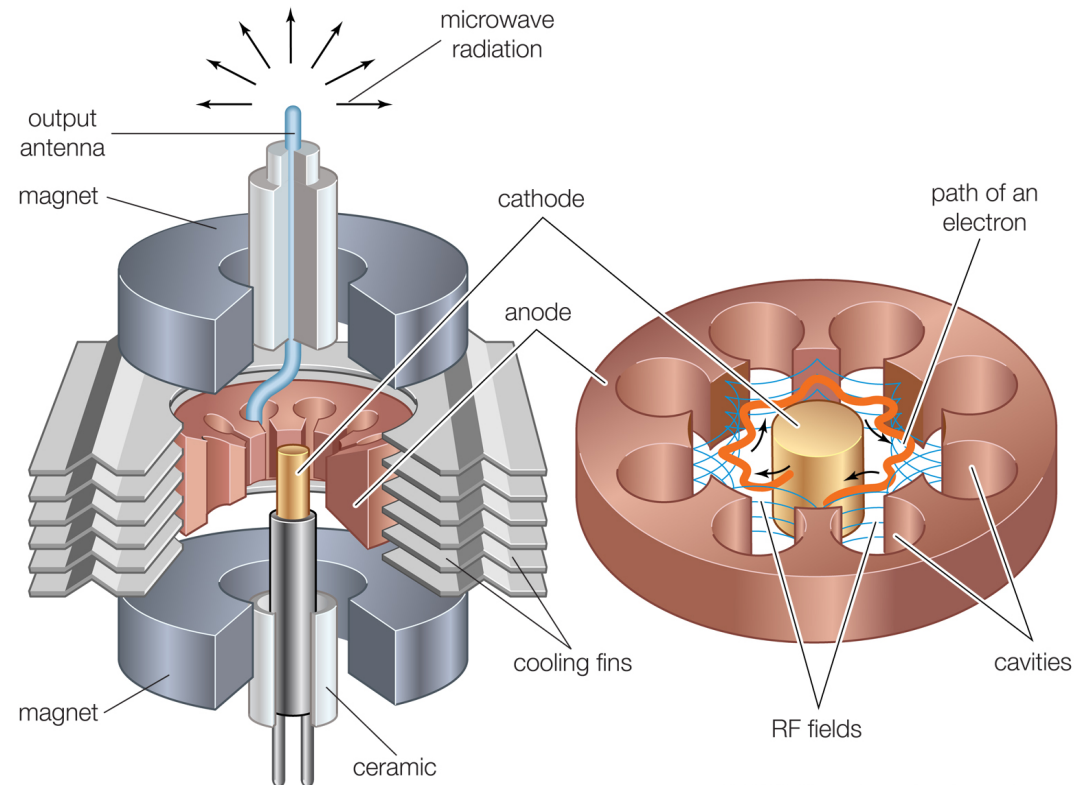
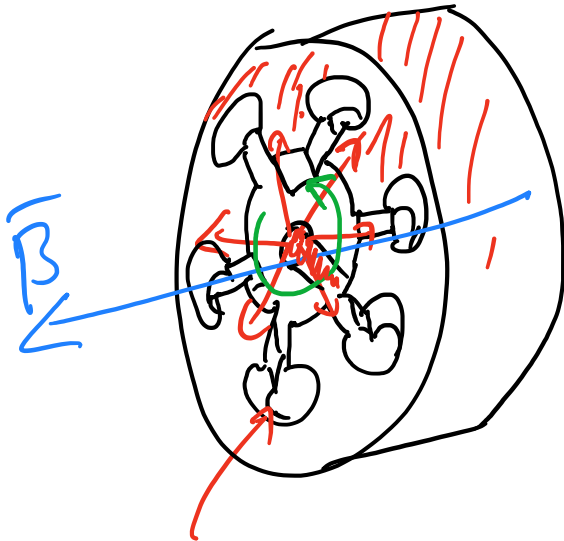
# Examples



Example of resonator: cavity magnetron source of 2.45 GHz radiation in microwave ovens (and WW2 radar)

$$\vec{F} = q\vec{v} \times \vec{B}$$

$\uparrow$   
 $-e$



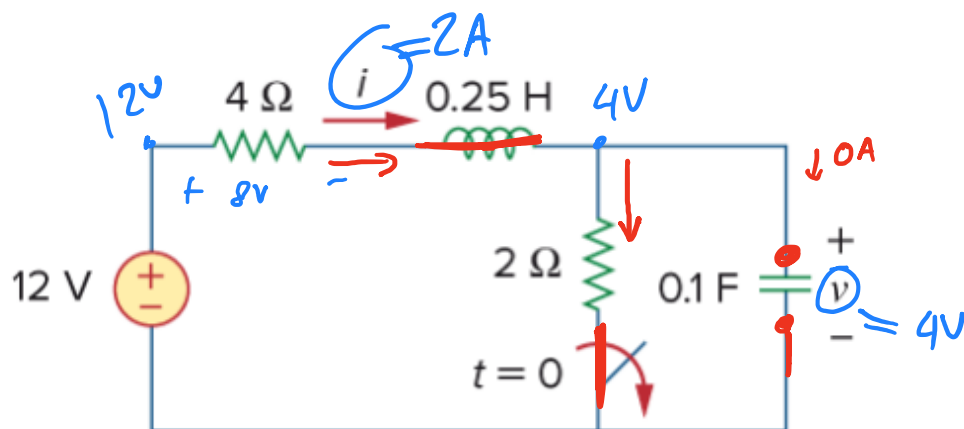
# Initial conditions

- We will need:  $v(0)$ ,  $i(0)$ ,  $dv(0)/dt$ ,  $di(0)/dt$ ,  $i(\infty)$ , and  $v(\infty)$ .
- Remember: use passive sign convention
- For capacitor:  $v(0^-) = v(0^+)$
- For inductor:  $i(0^-) = i(0^+)$

# Example 8.1

The switch in **Fig. 8.2** has been closed for a long time. It is open at  $t = 0$ . Find:

(a)  $i(0^+)$ ,  $v(0^+)$ , (b)  $di(0^+)/dt$ ,  $dv(0^+)/dt$ , (c)  $i(\infty)$ ,  $v(\infty)$ .

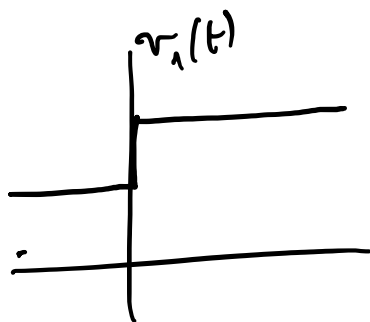


$$\frac{di(0^+)}{dt} \equiv \left. \frac{di(t)}{dt} \right|_{t=0^+}$$

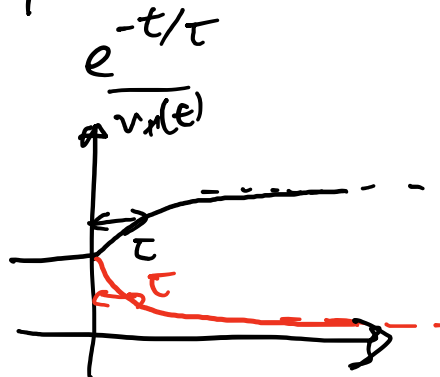
$$t=0^-: i=2A, v=4V, \frac{di}{dt}=0, \frac{dv}{dt}=0$$

$$t=0^+: i=2A, v=4V$$

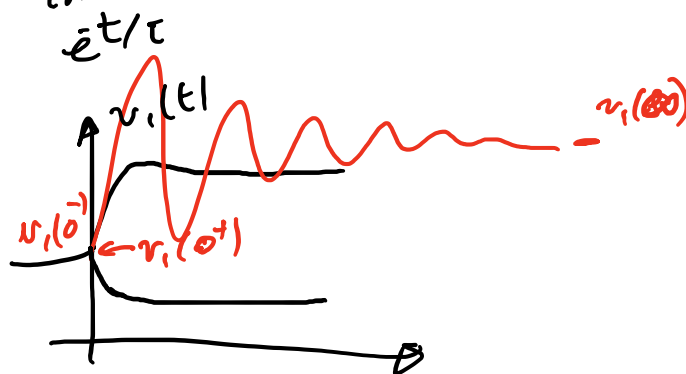
Only  $R$ 's



1st order



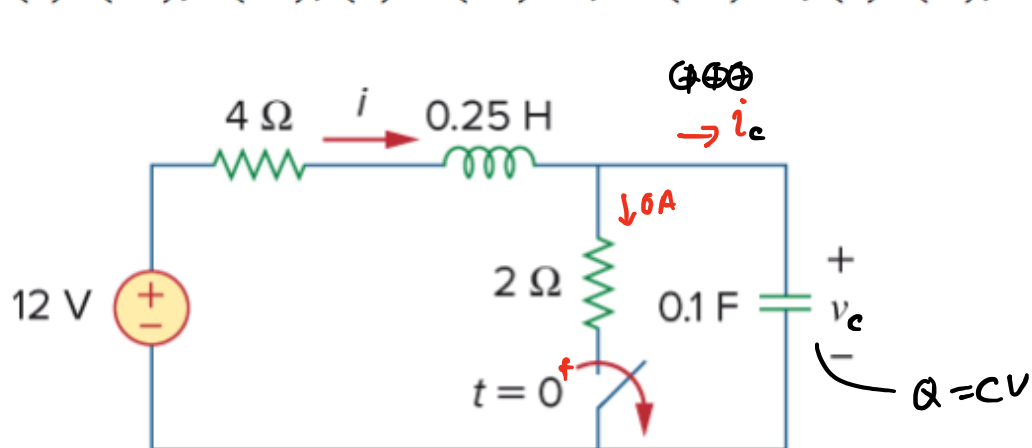
2nd order



## Example 8.1

The switch in **Fig. 8.2** has been closed for a long time. It is open at  $t = 0$ . Find:

(a)  $i(0^+)$ ,  $v(0^+)$ , (b)  $di(0^+)/dt$ ,  $dv(0^+)/dt$ , (c)  $i(\infty)$ ,  $v(\infty)$ .

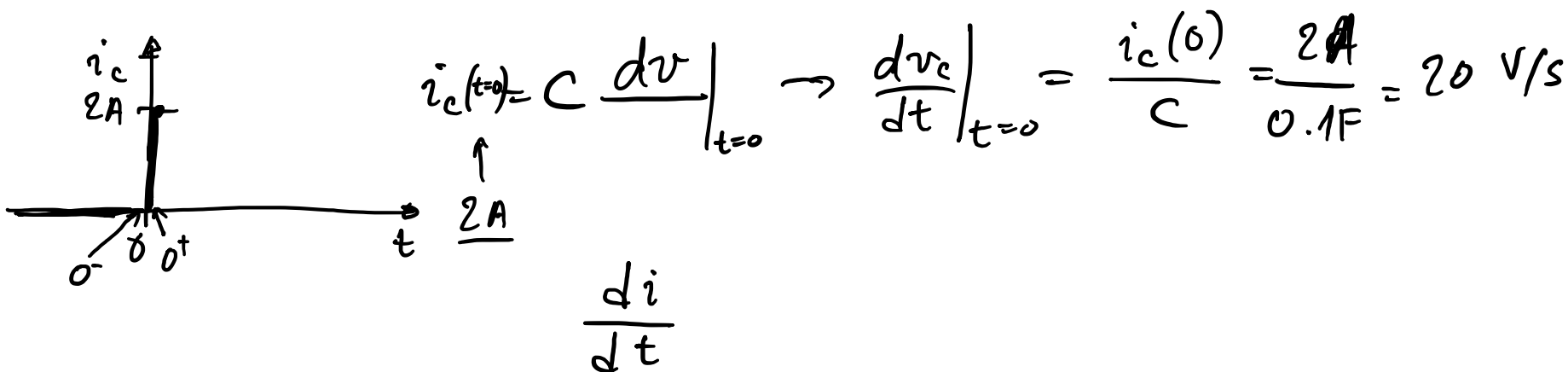


$$\frac{di(0^+)}{dt} \equiv \left. \frac{di(t)}{dt} \right|_{t=0^+}$$

$$t=0^-: i=2A, v=4V, \frac{di}{dt}=0, \frac{dv}{dt}=0$$

$$t=0^+: i=2A, v=4V$$

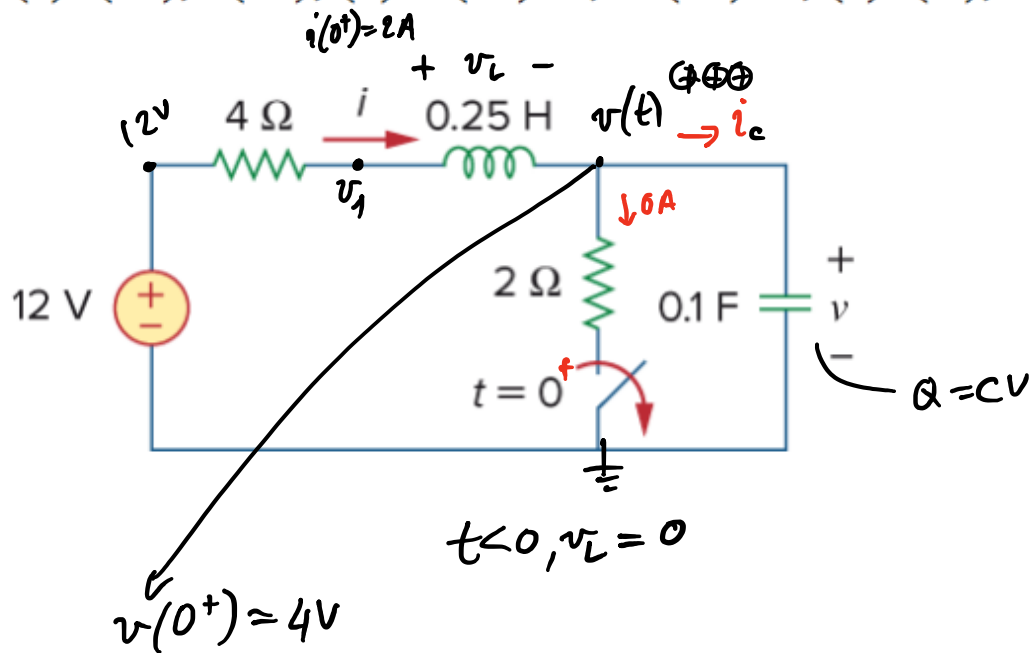
$$Q = CV$$



# Example 8.1

The switch in Fig. 8.2 has been closed for a long time. It is open at  $t = 0$ . Find:

(a)  $i(0^+)$ ,  $v(0^+)$ , (b)  $di(0^+)/dt$ ,  $dv(0^+)/dt$ , (c)  $i(\infty)$ ,  $v(\infty)$ .



$$\frac{di(0^+)}{dt} \equiv \left. \frac{di(t)}{dt} \right|_{t=0^+}$$

$$t=0^-: i=2A, v=4V, \frac{di}{dt}=0, \frac{dv}{dt}=0$$

$$t=0^+: i=2A, v=4V$$

$$t < 0: v_L = L \frac{di}{dt} = 0$$

$$t=0^+: v_L = 12V - 4V = 8V$$

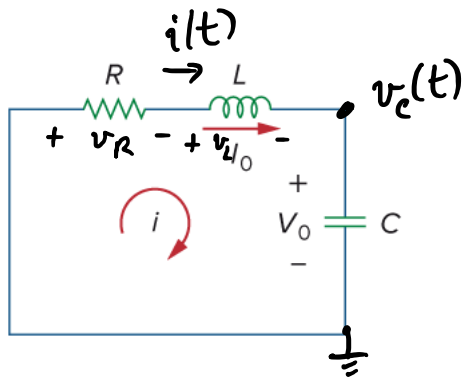
$$v_L = L \left. \frac{di}{dt} \right|_{t=0^+} = \frac{v_L}{L} = 0$$

$$= v_1 - v(t=0^+)$$

$$= 4V - 4V$$

# Source-free Series RLC Circuit

Give:  $v(0), i(0)$  at  $t=0$ . Find:  $v(t), i(t), t > 0$ .



$$\frac{d^2 i}{dt^2} + \frac{R}{L} \frac{di}{dt} + \frac{i}{LC} = 0$$

$$i = C \frac{dv_C}{dt} \rightarrow \frac{dv_C(t)}{dt} = \frac{i(t)}{C} \Rightarrow v_C(t) = \int_0^t \frac{1}{C} i(t') dt'$$

$$\begin{aligned} v_R(t) + v_L(t) + v_C(t) &= 0 \\ iR + L \frac{di}{dt} + \frac{1}{C} \int_0^t i(t') dt' &= 0 \end{aligned}$$

$$\frac{d}{dt} \left( iR + L \frac{di}{dt} + \frac{1}{C} \int_0^t i(t') dt' \right) = 0$$

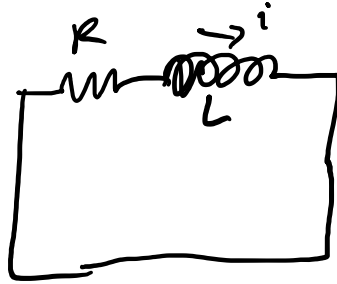
$$R \frac{di}{dt} + L \frac{d^2 i}{dt^2} + \frac{1}{C} i = 0$$

we need  $i(0)$   
 $i'(0) = \frac{v_L(0)}{L} \Rightarrow v_C \dots$

$$\frac{d^2}{dt^2} i + \frac{R}{L} \frac{di}{dt} + \frac{1}{LC} i = 0$$

$\frac{1}{LC} \rightarrow s^{-2} \rightarrow \frac{1}{\sqrt{LC}} \rightarrow s^{-1}$

$$L \frac{d}{dt} i + iR = 0$$



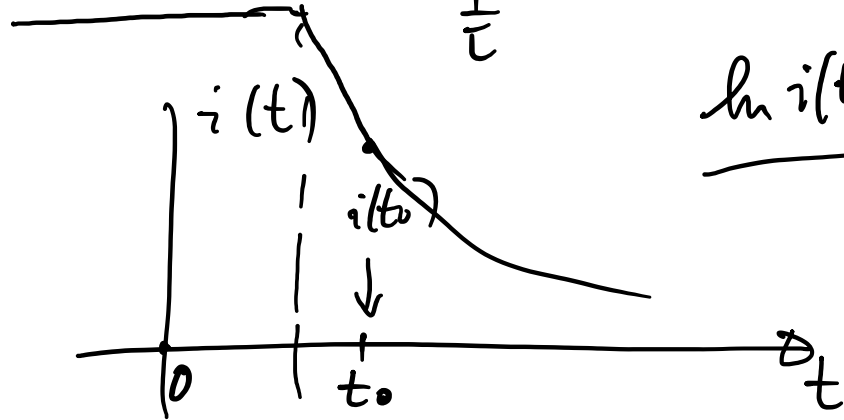
$$\frac{d}{dt} i + \frac{R}{L} i = 0$$

$$\frac{1}{i} \frac{d}{dt} i = -\frac{R}{L}$$

$$\frac{d}{dt} \ln i = -\frac{R}{L}$$

$$\ln i = -\frac{R}{L} \int dt = -\frac{R}{L} t$$

$$i(t) = e^{-\frac{R}{L} t}$$



$$\int_0^t \frac{d}{dt'} \ln i(t') dt' =$$

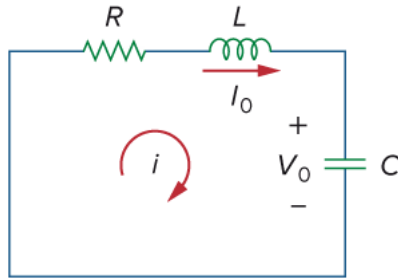
$$\ln i(t') \Big|_{t'=0}^{t'=t} = - \frac{t'=t}{t'=0} = -\frac{R}{L} (t-0)$$

$$\ln i(t) - \ln i(0) = -\frac{R}{L} t \Rightarrow \frac{i(t)}{i(0)} = e^{-\frac{R}{L} t}$$

$$\ln i(t) - \ln i(t_0) = -\frac{R}{L} (t - t_0)$$

$$\boxed{\frac{i(t)}{i(t_0)} = e^{-\frac{R}{L} (t - t_0)}}$$

# Source-free Series RLC Circuit



$$\frac{d^2}{dt^2} i + \frac{R}{L} \frac{di}{dt} + \frac{1}{LC} i = 0$$

$$s^2 \cancel{Ae^{st}} + \frac{R}{L} s \cancel{Ae^{st}} + \frac{1}{LC} \cancel{Ae^{st}} = 0$$

$$\boxed{s^2 + \frac{R}{L}s + \frac{1}{LC} = 0}$$

$$s = -\frac{1}{\tau}$$

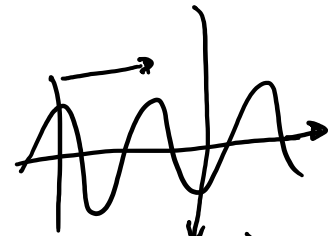
$$e^{-\frac{1}{\tau}t}$$

$$s = j\omega_0$$

$$(j\omega_0)t$$

$$= \frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2}$$

freq (rad/s)



$$\cos(\omega_0 t)$$

$$\frac{di}{dt} = sAe^{st}$$

$$\frac{d^2 i}{dt^2} = s^2 Ae^{st}$$

$$\boxed{i = Ae^{st}}$$

$$\boxed{s^2 + \frac{R}{L}s + \frac{1}{LC} = 0}$$

$$s_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}}$$

$$s_2 = -\frac{R}{2L} - \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}}$$

$s_1$  and  $s_2$  : natural frequencies, measured in nepers per second (Np/s).

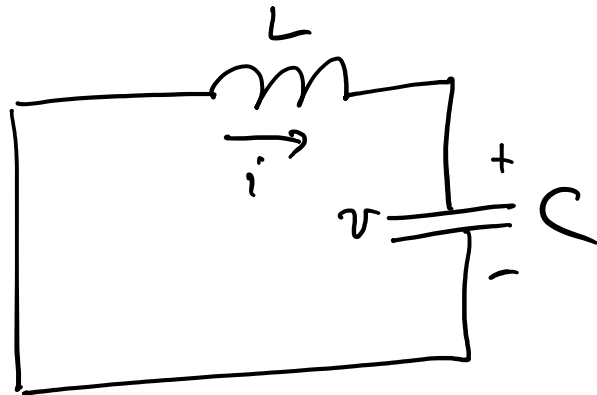
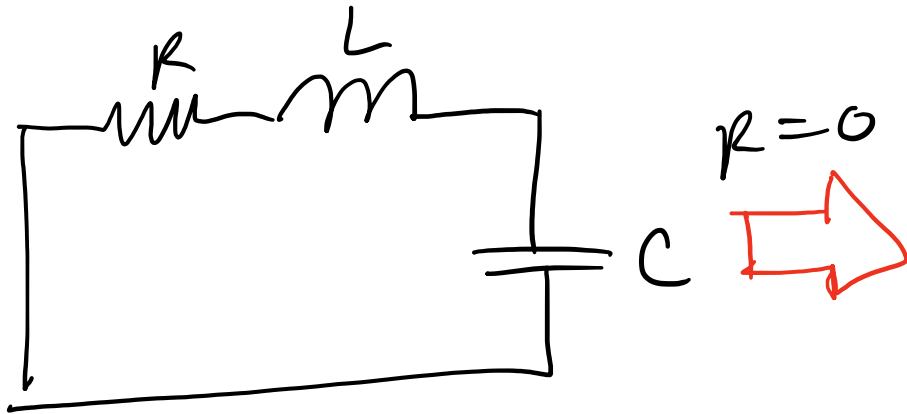
$\omega_0$  : resonant frequency or strictly as the undamped natural frequency, expressed in radians per second (rad/s)

$$Ax^2 + Bx + C = 0$$

$$x = \frac{-B \pm \sqrt{B^2 - 4AC}}{2A}$$

$$s = \frac{-\frac{R}{L} \pm \sqrt{\left(\frac{R}{L}\right)^2 - 4\frac{1}{LC}}}{2}$$

$$\boxed{s = -\frac{R}{2L} \pm \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}}}$$



$$s = \frac{-R}{2L} \pm \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}}$$

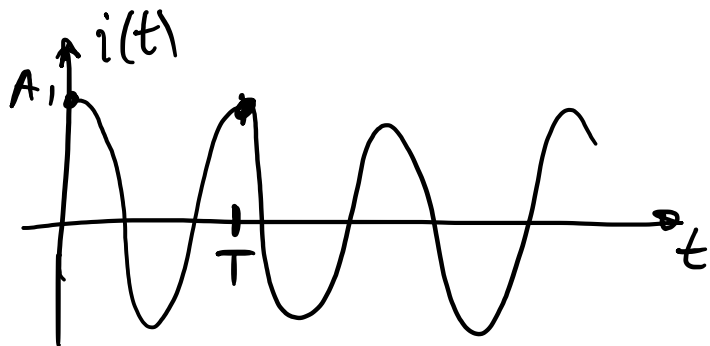
$$s = 0 \pm \sqrt{0 - \frac{1}{LC}}$$

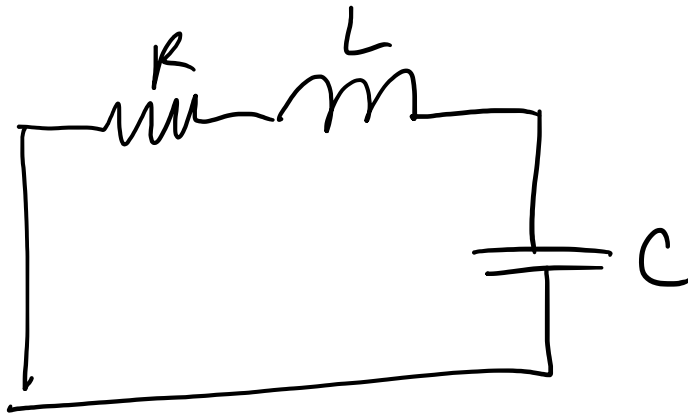
$$s_{1,2} = \pm j \frac{1}{\sqrt{LC}} = \pm j\omega_0$$

$$i(t) = A_1 e^{s_1 t} + A_2 e^{s_2 t} = A_1 e^{j\omega_0 t} + A_2 e^{-j\omega_0 t}$$

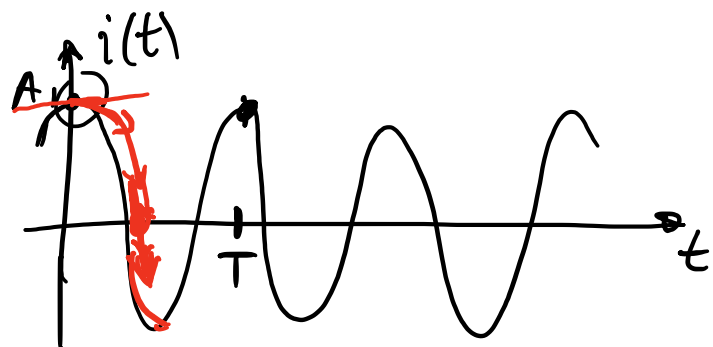
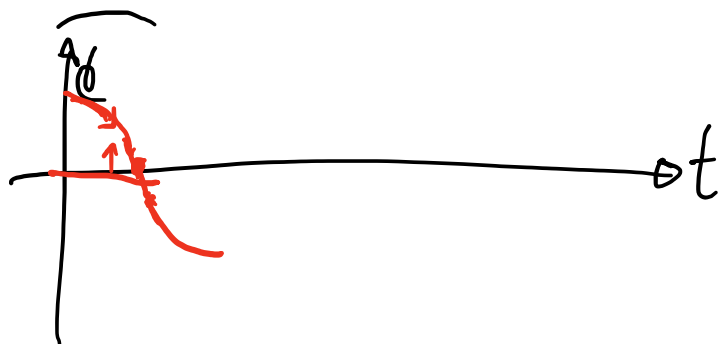
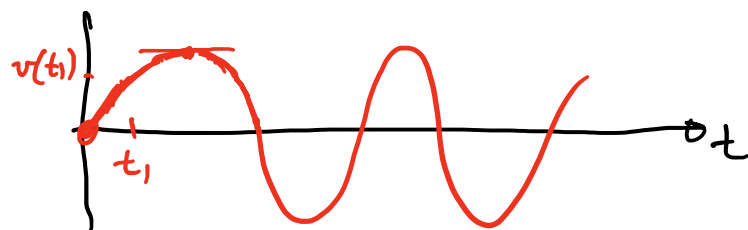
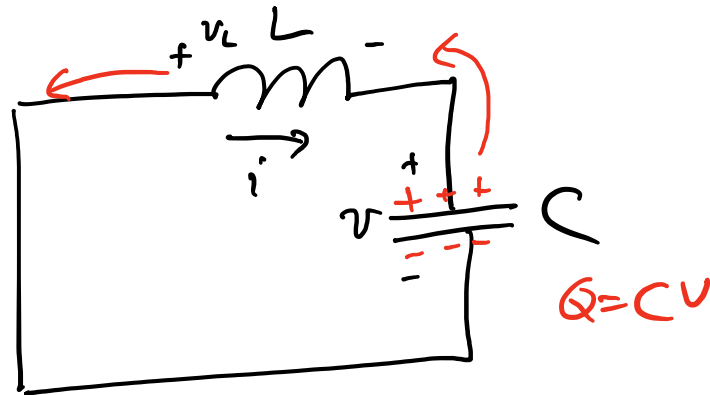
$$= A_1 (e^{j\omega_0 t} + e^{-j\omega_0 t}) = \underline{A_1 \cos(\omega_0 t)}$$

$$T = \frac{2\pi}{\omega_0}$$





$$R=0$$

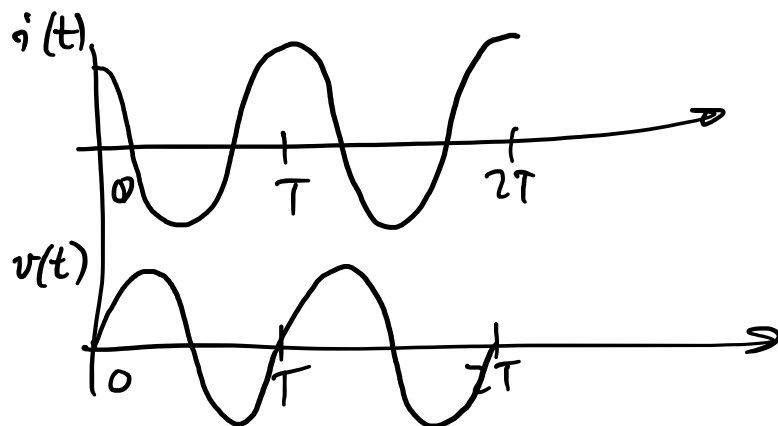


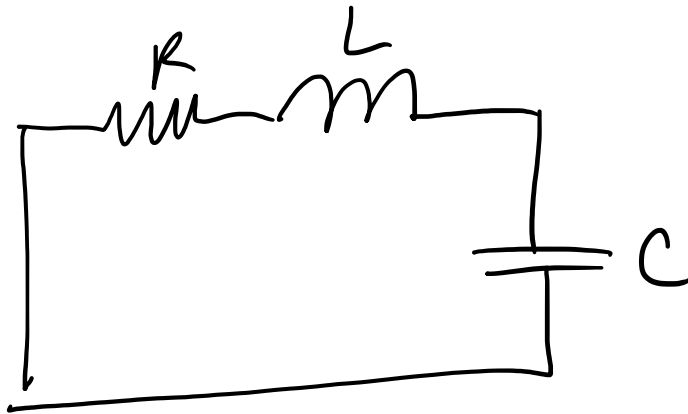
$$-v = +v_L = L \frac{di}{dt} \quad i = C \frac{dv}{dt}$$

$$\frac{di}{dt} = -\frac{v}{L}$$

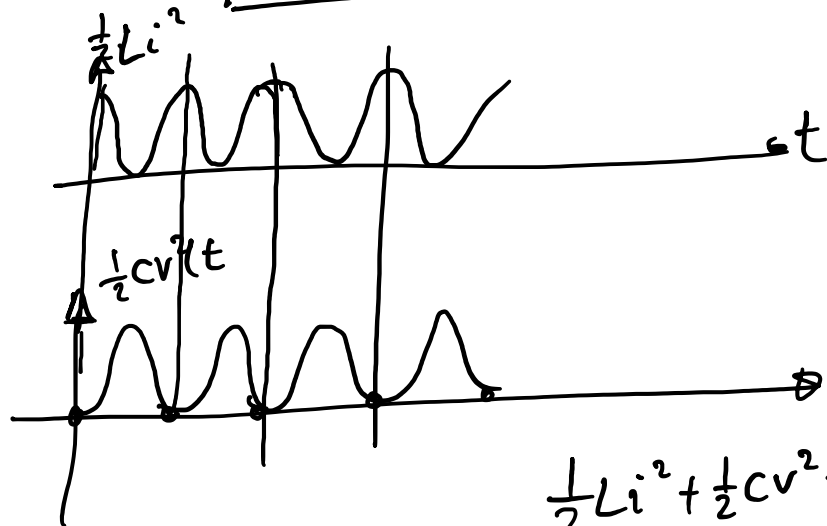
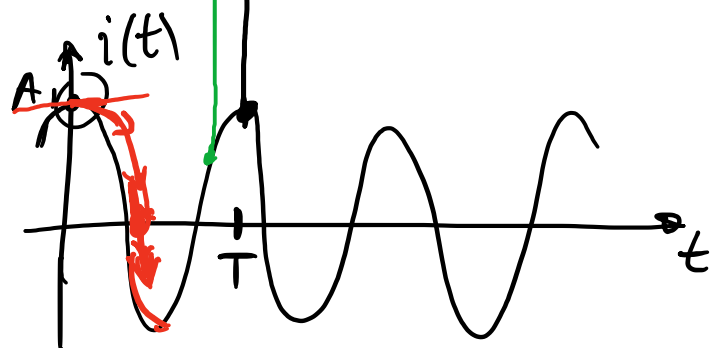
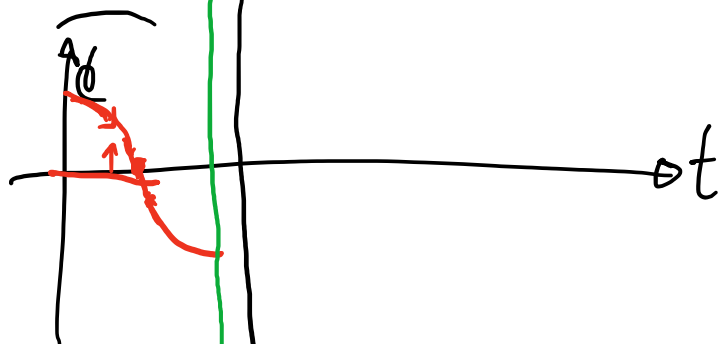
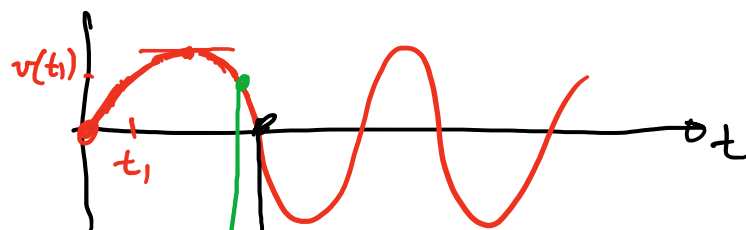
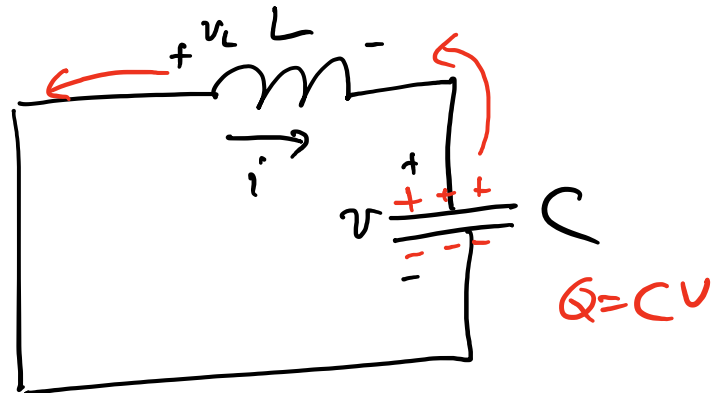
$$= 0 \text{ at } t=0$$

$$\left. \frac{di}{dt} \right|_{t=t_1} = -\frac{v(t_1)}{L} < 0$$

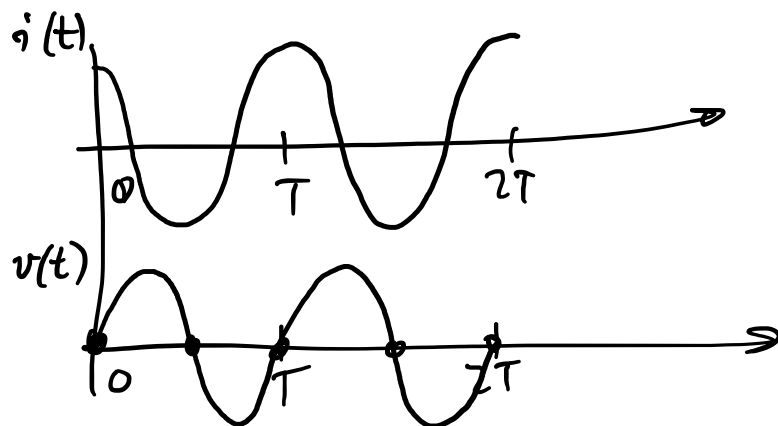




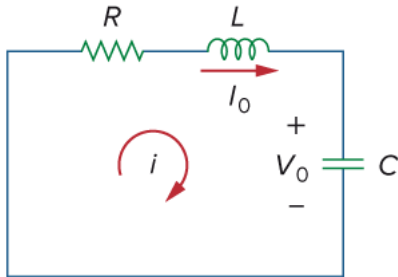
$$R=0$$



$$\frac{1}{2}Li^2 + \frac{1}{2}Cv^2 = \text{const.}$$



# Source-free Series RLC Circuit



$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$$

$$s^2 + 2\alpha s + \omega_0^2 = 0$$

$$i_1 = A_1 e^{s_1 t}, \quad i_2 = A_2 e^{s_2 t}$$

$$i(t) = A_1 e^{s_1 t} + A_2 e^{s_2 t}$$

$$s_1 = -\alpha + \sqrt{\alpha^2 - \omega_0^2}, \quad s_2 = -\alpha - \sqrt{\alpha^2 - \omega_0^2}$$

where

$$\alpha = \frac{R}{2L}, \quad \omega_0 = \frac{1}{\sqrt{LC}}$$

↑  
resonant freq.

We can infer 3 types of soln:

- 1) If  $\alpha > \omega_0$ , we have the *overdamped* case.
- 2) If  $\alpha = \omega_0$ , we have the *critically damped* case.
- 3) If  $\alpha < \omega_0$ , we have the *underdamped* case.

# Source-free Series RLC Circuit

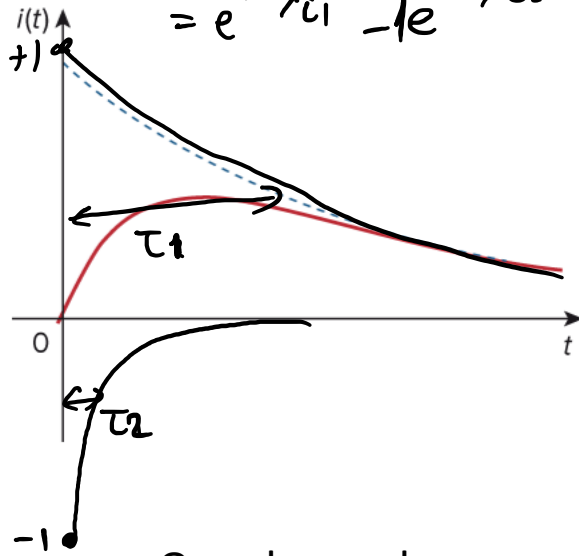
We can infer 3 types of soln:

- 1) If  $\alpha > \omega_0$ , we have the *overdamped* case.
- 2) If  $\alpha = \omega_0$ , we have the *critically damped* case.
- 3) If  $\alpha < \omega_0$ , we have the *underdamped* case.

$$s = -\alpha \pm \Delta\alpha$$

$$i(t) = e^{-\alpha_1 t} - e^{-\alpha_2 t}$$

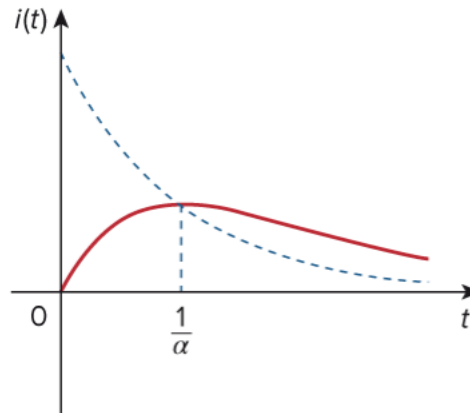
$$= e^{-t/\tau_1} - e^{-t/\tau_2}$$



Overdamped

$$s = -\alpha \pm 0$$

$$i(t) = A_1 e^{-\alpha t} + A_2 t e^{-\alpha t}$$



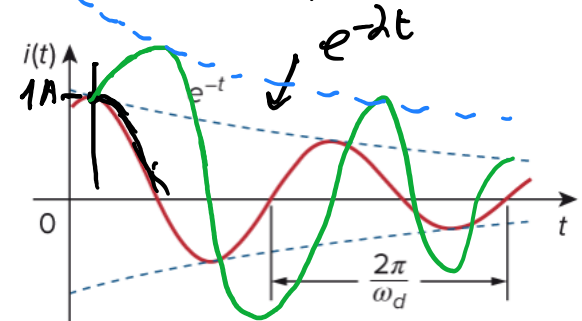
Critically damped

$$s = -\alpha \pm j\omega_0$$

$$i(t) = A_1 e^{(-\alpha + j\omega_0)t} + A_2 e^{(-\alpha - j\omega_0)t}$$

$$= A_1 2 \cos(\omega_0 t) e^{-\alpha t}$$

if  $A_1 = A_2$



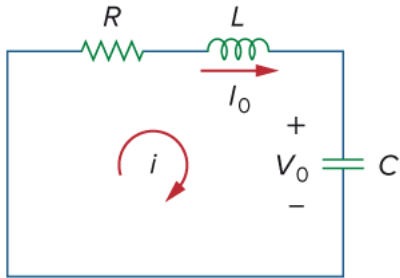
$$i(0) = 1A, v(0) = 0$$

$$i(0) = 1A, v(0) = 1V$$

Underdamped

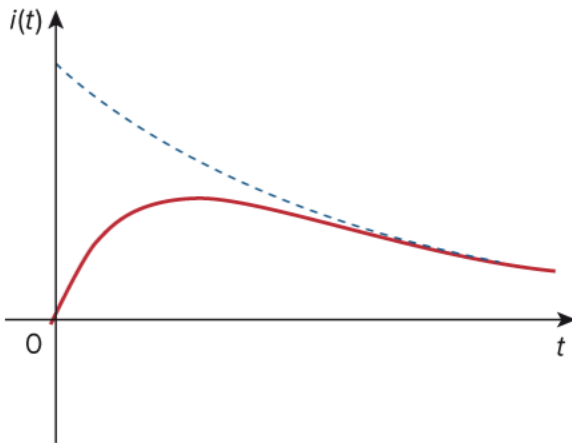
# Overdamped case

$$\alpha > \omega_0 \quad \Rightarrow \quad C > 4L/R^2$$

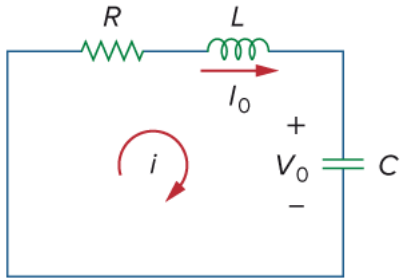


$$i(t) = A_1 e^{s_1 t} + A_2 e^{s_2 t}$$

$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$$



# Critically damped case



$$\alpha = \omega_0 \Rightarrow C = 4L/R^2$$

$$s_1 = s_2 = -\alpha = -\frac{R}{2L}$$

$$i(t) = A_1 e^{-\alpha t} + A_2 e^{-\alpha t} = A_3 e^{-\alpha t}$$

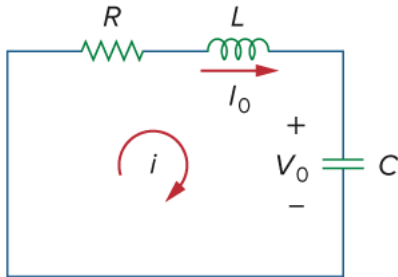
Assumption of exponential solution  
wrong for critical damping case!

$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$$

$$\frac{d^2 i}{dt^2} + \frac{R}{L} \frac{di}{dt} + \frac{i}{LC} = 0 \Rightarrow \frac{d^2 i}{dt^2} + 2\alpha \frac{di}{dt} + \alpha^2 i = 0$$

# Critically damped case

$$\alpha = \omega_0 \rightarrow C = 4L/R^2$$



$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$$

$$\frac{d^2 i}{dt^2} + 2\alpha \frac{di}{dt} + \alpha^2 i = 0$$

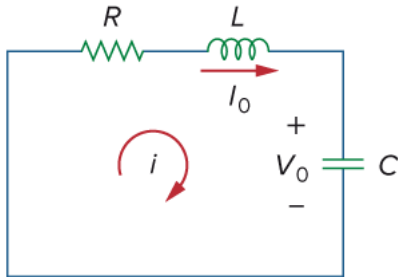
$$\frac{d}{dt} \left( \frac{di}{dt} + \alpha i \right) + \alpha \left( \frac{di}{dt} + \alpha i \right) = 0$$

$$f = \frac{di}{dt} + \alpha i$$

$$\frac{df}{dt} + \alpha f = 0$$

# Critically damped case

$$\alpha = \omega_0 \rightarrow C = 4L/R^2$$



$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$$

$$\frac{df}{dt} + \alpha f = 0$$

$$\frac{di}{dt} + \alpha i = A_1 e^{-\alpha t}$$

$$\frac{d}{dt}(e^{\alpha t} i) = A_1$$

$$e^{\alpha t} i = A_1 t + A_2$$

$$i = (A_1 t + A_2) e^{-\alpha t}$$

